

Acoustic Willis meta-atom beyond the bounds of passivity and reciprocity

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Willis metamaterial enables exotic manipulations of acoustic waves with a precise combination of bulk modulus, mass density, and Willis parameters. While the realization of unrestricted and completely decoupled constitutive parameters would extend the horizon of future applications, the restriction of passivity and reciprocity dictate a hard bound in the values of achievable polarizabilities and correlations between them. Here, we break the bound of passivity and reciprocity by instituting a basis and independent kernel for each constitutive polarization in a virtualized metamaterial platform, active metamaterials realizing artificial polarization with the digital convolution. We demonstrate decoupled control of all four constitutive parameters in a nonreciprocal regime, at the same time achieving values of polarizabilities beyond the passivity limit. Broadband, flat-response nonreciprocal Willis coupling is also demonstrated with analytically designed causal frequency dispersion. Our approach will be useful for nonreciprocal wave manipulation and communication for broadband operation.

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Metamaterials enable novel applications by providing unconventional wave properties in many wave systems^{1–3}. In acoustic systems, unusual wave parameters such as extremely high⁴, negative^{5,6}, or zero bulk moduli and mass densities^{7,8} have now also become a reality, enabling exotic applications such as superfocusing^{9,10}, extraordinary diffraction¹¹, and acoustic cloaking^{12–14}. Recently, with the experimental realization of the Willis coupling^{7,15–21} as an acoustic duality of bianisotropy in electromagnetic waves²², the scope of acoustic metamaterials, also known as acoustic Willis metamaterials in extended perspective, has been greatly extended by allowing coupling between pressure and velocity fields. Designer Willis parameter combined with a precisely tuned bulk modulus and mass density, acoustic metamaterials now enable applications that were impossible before, such as the separate control of reflected and transmitted waves⁷, diffraction-free metasurfaces²³, bianisotropic nihility²⁴, and metagratings^{25–27}.

Despite the great successes achieved thus far, the full potential of Willis metamaterials has not been fully explored. While the unrestricted and decoupled control of all four constitutive parameters would be ideal, yet the passivity and reciprocity in relation to the metamaterial structure command stringent bound in their values and correlations. In order to realize acoustic properties that go beyond conventional metamaterials, it has been proposed that electroacoustic metamaterials utilizing transducer arrays achieve control of wave parameters such as mass density^{28–30}, bulk modulus^{30–32}, and acoustic Willis coupling parameters^{18,19,33–35} and also enable the manipulation of acoustic waves^{34–37}. Although it is clear that active components could provide total power larger than the input, and while single-polarization control of nonreciprocal and broadband active Willis coupling has been demonstrated in acoustic^{18,35} and elastic waves¹⁹, the breaking of the passive Willis bound and nonreciprocal control of all four constitutive parameters has not been realized as it requires precise control of balance on realizing stable resonance.

In this work, we break the bound of passivity and the conditions for nonreciprocity, extending the boundary of Willis metamaterial to the causality limit. We analyze the passivity condition and its implications to the correlation between four constitutive parameters and then institute the condition of overcoming these limitations based on the recently developed

concept of virtualized meta-atoms³¹. By employing a representation basis for each polarization and then applying independent scattering kernel functions, we demonstrate the breaking of passive bounds of all four polarizabilities and correlations between them to achieve a full nonreciprocity. As an example of the wide-open flexibility in the manipulation of polarizabilities, we further realize Willis metamaterial with broadband and flat bianisotropy, in both the purely reciprocal and purely non-reciprocal regimes, from the dispersion curve analytically constructed via the inverse design method.

Results

Polarization process based on parity symmetry. To study the condition of passivity and its implications to the correlation between four constitutive parameters, here we define a scattering matrix \mathbf{S} having parity symmetry¹⁷ in the one-dimensional system, as depicted in Fig. 1, for incident (a) and scattered (b) waves propagating in the forward (+) and backward (–) directions, which are decomposed into components of even (e) and odd (o) parities: $a_e = (a_+ + a_-)/2$, $a_o = (a_+ - a_-)/2$, $b_e = (b_+ + b_-)/2$, and $b_o = (b_+ - b_-)/2$. The couplings between the incident fields and scattered fields are then written as $(b_e \ b_o)^T = \mathbf{S} (a_e \ a_o)^T$, where the superscript T is transpose operator with the scattering matrix \mathbf{S} defined as

$$\begin{pmatrix} s_{ee} & s_{eo} \\ s_{oe} & s_{oo} \end{pmatrix} = \frac{1}{2} \begin{pmatrix} t_+ + t_- + r_+ + r_- - 2 & t_+ - t_- + r_+ - r_- \\ t_+ - t_- - r_+ + r_- & t_+ + t_- - r_+ - r_- - 2 \end{pmatrix}, \quad (1)$$

where r and t are the reflection and transmission coefficients, respectively. Then the relation between the scattering matrix \mathbf{S} and the normalized polarizability α becomes:

$$\begin{pmatrix} \alpha_{pp} & \alpha_{pv} \\ \alpha_{vp} & \alpha_{vv} \end{pmatrix} = \frac{1}{ik_0} \begin{pmatrix} s_{ee} & s_{eo} \\ s_{oe} & s_{oo} \end{pmatrix}, \quad (2)$$

where k_0 is the free space wavenumber (see Supplementary Note S1 for the derivation of and detailed expressions for each element). In this representation, the diagonal terms relating even-incident to even-scattering components (s_{ee}) and odd-incident to odd-scattering components (s_{oo}) correspond to the inverse bulk modulus and mass density, respectively, while the off-diagonal

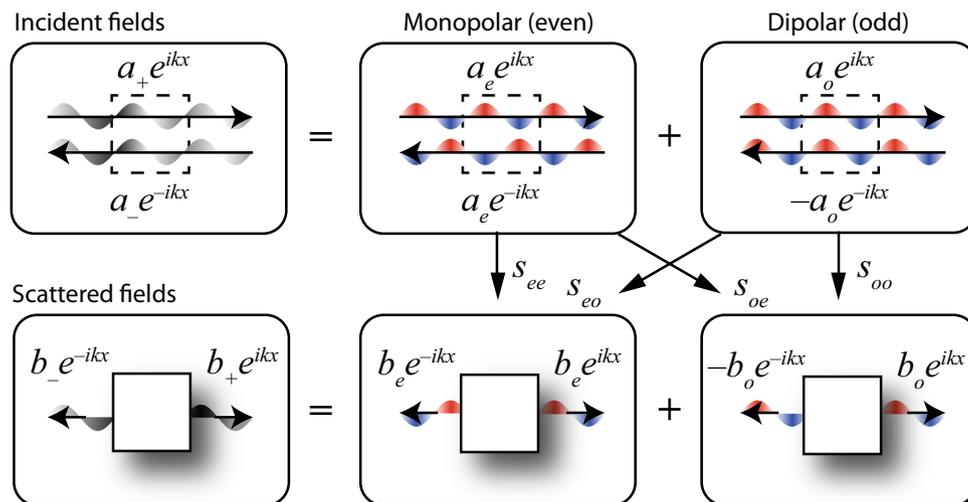


Fig. 1 Scattering matrix \mathbf{S} based on coupling symmetry. The definition of the scattering matrix \mathbf{S} in a one-dimensional system. The incident (a) and resultant scattered fields (b) propagating in the forward (+) and backward (–) directions are decomposed into even (e) and odd (o) components where k is the wavenumber. The even-to-even and odd-to-odd scattering parameters correspond to the inverse bulk modulus and density, respectively, while the even-to-odd and odd-to-even scattering parameters are the Willis parameters.

components coupling even to odd components (s_{oe}) and odd to even components (s_{eo}) are the acoustic bianisotropy or Willis coupling parameters.

Willis coupling beyond the passivity bound. Following the derivation in Li et al.²⁵ for two- and three-dimensional systems, the bound and correlation for Willis parameters and other polarizabilities in the one-dimensional passive system is dictated by the following two inequalities (see Supplementary Note S2 for the derivation):

$$\begin{aligned} |k_0 \alpha_{vp}|^2 + |1 + ik_0 \alpha_{pp}|^2 &\leq 1, \\ |k_0 \alpha_{pv}|^2 + |1 + ik_0 \alpha_{vv}|^2 &\leq 1. \end{aligned} \quad (3)$$

Thus, the maximum bound of the Willis coupling is given by $|\alpha_{vp}| (|\alpha_{pv}|) \leq k_0^{-1}$, where the equality is satisfied *only* when $\alpha_{pp} (\alpha_{vv}) = ik_0^{-1}$. If the systems are strictly reciprocal, i.e., $t_+ = t_-$, then the inequalities in Eq. (3) are reduced to a single inequality equation of $|r_+ - r_-| \leq 2$, where the maximum Willis coupling is achieved only at the stringent condition of $t_+ = t_- = 0$ and $r_+ = -r_- = e^{i\varphi}$, where φ is the arbitrary real number. It is noted that the passivity condition not only restricts the maximum value of Willis couplings $|\alpha_{vp}|$ and $|\alpha_{pv}|$, but also bounds the achievable diagonal polarizabilities α_{pp} by $|1 + ik_0 \alpha_{pp}| \leq 1$ at $\alpha_{vp} = 0$, and α_{vv} by $|1 + ik_0 \alpha_{vv}| \leq 1$ at $\alpha_{pv} = 0$.

To break the correlation in Eq. (3) and to realize all the polarization responses independently without bound, it is needed to unharness the condition of $t_+ = t_-$ and $|t_{+(-)}|^2 + |r_{+(-)}|^2 \leq 1$, in our case employing a platform of a virtualized meta-atom - which can feed energy to the system and directly mold scattering parameters with designer convolution functions³¹. As depicted in

Fig. 2, the microprocessor returns output values to two speakers (S_i) from the detected signals of two microphones (M_j) by means of the programmed convolution kernels (\tilde{Y}_{ij}). I.e., the output voltages of the sources are calculated in the time domain as follows:

$$S_i(t) = -\partial_t^2 \left[\tilde{Y}_{ij}(t) * M_j(t) \right], \quad (4)$$

where $*$ is the convolution operator, and the subscripts $i, j = 1$ or 2 are the labels of the speakers and microphones.

In the frequency domain, the entire operation is summarized as $S_i(\omega) = Y_{ij}(\omega) M_j(\omega)$ where $Y_{ij}(\omega) = \omega^2 \tilde{Y}_{ij}(\omega)$. To achieve a connection between the speaker output S_i and the microphone-detected signal M_j , similar to the polarization process in Eq. (1), we decompose the convolution kernel Y_{ij} by introducing a basis of convolution matrices:

$$\mathbf{e}_{ee} = \frac{1}{2} \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix}, \mathbf{e}_{eo} = \frac{1}{2} \begin{pmatrix} -1 & 1 \\ -1 & 1 \end{pmatrix}, \mathbf{e}_{oe} = \frac{1}{2} \begin{pmatrix} -1 & -1 \\ 1 & 1 \end{pmatrix}, \mathbf{e}_{oo} = \frac{1}{2} \begin{pmatrix} 1 & -1 \\ -1 & 1 \end{pmatrix}, \quad (5)$$

which satisfy $\mathbf{Y} = Y_{ee} \mathbf{e}_{ee} + Y_{eo} \mathbf{e}_{eo} + Y_{oe} \mathbf{e}_{oe} + Y_{oo} \mathbf{e}_{oo}$. The relation between speaker outputs and microphone signals can then be rewritten in terms of the symmetric and antisymmetric components of the speakers and microphones as follows:

$$\begin{pmatrix} S_1 + S_2 \\ S_2 - S_1 \end{pmatrix} = \begin{pmatrix} Y_{ee} & Y_{eo} \\ Y_{oe} & Y_{oo} \end{pmatrix} \begin{pmatrix} M_1 + M_2 \\ M_2 - M_1 \end{pmatrix}. \quad (6)$$

It is important to note that this representation in terms of a constitutive parameter basis \mathbf{e}_{ij} and coefficients Y_{ij} , instead of usual S-parameters¹⁸, enables independent access toward each of the four polarizability parameters, including the two Willis parameters. In our meta-atom configuration shown in Fig. 2, with excitation (i.e., $\mathbf{Y} = \sum Y_{ij} \mathbf{e}_{ij}$) in the weak excitation limit ($Y_{ij} \ll 1$),

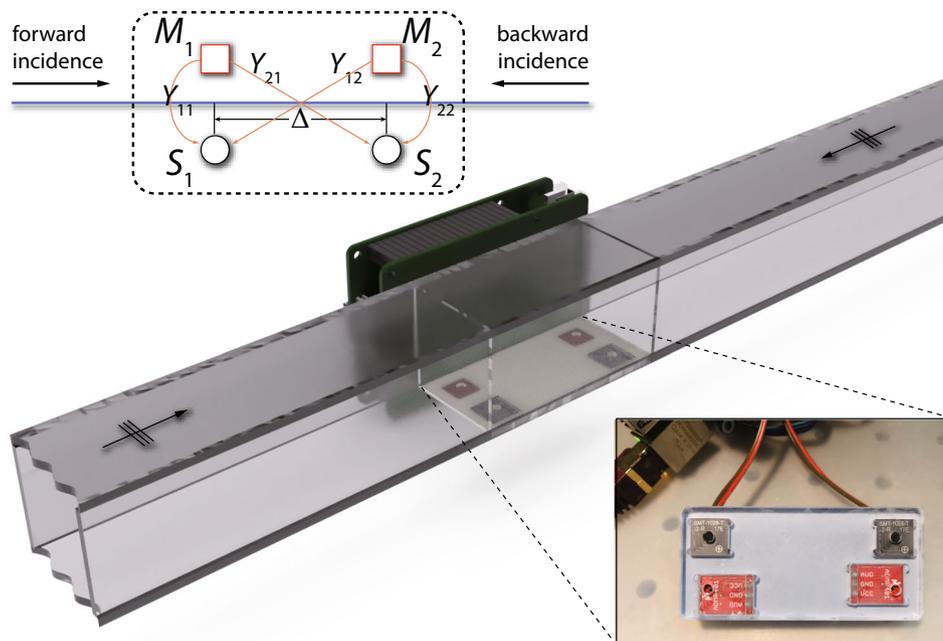


Fig. 2 **Bianisotropic virtualized metamaterial.** A virtualized metamaterial consisting of two microphones and two speakers connected to a microprocessor embedded in the cover of an acoustic waveguide. The bottom-right inset shows a photograph of a transducer module of a virtualized meta-atom. The top inset shows an analytical representation of the virtualized metamaterial atom: two microphones (M_j) are convolved with a 2×2 matrix (Y_{ij}) returning signals to fire at the two speakers (S_i) as secondary radiation. The distances between the two speakers and microphones are equally set to $\Delta = 50$ mm. The scattering property of the meta-atom is tested by incident waves coming from the forward and backward directions to identify all 4 scattering parameters (r_+ , t_+ , r_- , and t_-).

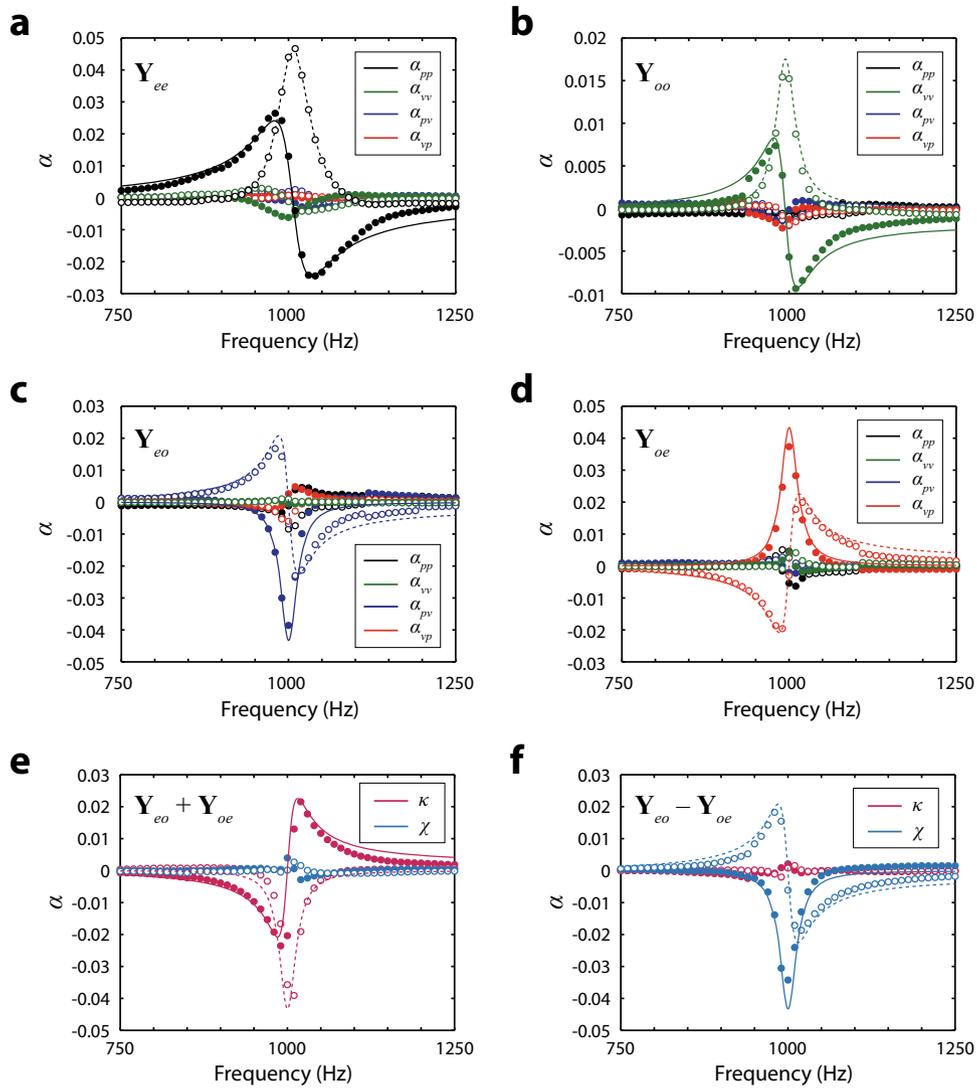


Fig. 3 Decoupled excitation of polarization components. **a-d** Virtualized Willis metamaterial for the four basis convolution matrices \mathbf{e}_{ee} (**a**), \mathbf{e}_{oo} (**b**), \mathbf{e}_{eo} (**c**), and \mathbf{e}_{oe} (**d**) with the same Lorentzian convolution kernel Y_0 , where the model parameters are set to $a = 2\pi \times 15$, $\theta = -\pi/2$, $\gamma = 2\pi \times 15$ Hz, and $\omega_0 = 2\pi \times 1.0$ kHz. The polarizabilities α_{pp} , α_{vv} , α_{pv} , and α_{vp} are depicted in black, green, blue and red, respectively, with solid/empty symbols representing the real/imaginary parts of the experimental results. The corresponding theoretical models are plotted with solid/dashed lines for the real/imaginary parts. **e, f** Purely reciprocal and purely nonreciprocal Willis couplings $\kappa = i(\alpha_{pv} - \alpha_{vp})/2$ and $\chi = (\alpha_{pv} + \alpha_{vp})/2$, realized with a balanced \mathbf{e}_{oe} and \mathbf{e}_{eo} . The purely reciprocal Willis coupling satisfying $\alpha_{vp} = -\alpha_{pv}$ is demonstrated by their summation, i.e., $\mathbf{Y} = Y_0(\mathbf{e}_{eo} + \mathbf{e}_{oe})$ (**e**), and the purely nonreciprocal term satisfying $\alpha_{vp} = \alpha_{pv}$ is demonstrated by subtracting the two basis convolution matrices, $\mathbf{Y} = Y_0(\mathbf{e}_{eo} - \mathbf{e}_{oe})$ (**f**).

the polarizability parameters can then be written as follows (see Supplementary Note S3):

$$\begin{aligned}
 \alpha_{pp} &= -4ik_0^{-1} \cos(k_0\Delta/2)^2 Y_{ee}, \\
 \alpha_{pv} &= 2k_0^{-1} \sin(k_0\Delta) Y_{eo}, \\
 \alpha_{vp} &= -2k_0^{-1} \sin(k_0\Delta) Y_{oe}, \\
 \alpha_{vv} &= -4ik_0^{-1} \sin(k_0\Delta/2)^2 Y_{oo},
 \end{aligned}
 \tag{7}$$

where Δ is the distance between the two speakers (and microphones). Therefore, decoupled control or balancing among all acoustic wave parameters can be realized with analytically constructed kernels Y_{ij} of the desired design. It is noted that the unabridged Eq. (S23) is analogous to the effective medium theory expression³⁸ for composite scatterers, however critically, except the independently addressable kernels Y_{oe} and Y_{eo} in our realization.

Figure 3 shows an experimental demonstration of the selective excitation of each polarizability parameter in the virtual Willis metamaterial. We set the program to have one of the basis convolution matrices \mathbf{e}_{ee} , \mathbf{e}_{eo} , \mathbf{e}_{oe} , and \mathbf{e}_{oo} given in Eq. (5) with the Lorentzian-form coefficient $\tilde{Y}_0(\omega)$. For time-domain microprocessor signal processing, $\tilde{Y}_0(\omega)$ can then be implemented as the following impulse response function:

$$\tilde{y}_0(t) = \frac{a}{\omega_0^2} \sin(\omega_0 t + \theta) e^{-\gamma t} u(t),
 \tag{8}$$

where $u(t)$ is the Heaviside step function, a is the total scaling factor, ω_0 is the resonance frequency, θ is the phase, and γ is the resonance bandwidth. Figure 3a, b shows the Lorentzian polarizations experimentally realized with even-to-even \mathbf{e}_{ee} and odd-to-odd \mathbf{e}_{oo} excitations, which are responsible for inverse bulk modulus and mass density, respectively, and Fig. 3c, d shows the implementation of bianisotropy achieved with odd-to-even \mathbf{e}_{eo} and even-to-odd \mathbf{e}_{oe} convolutions. Each polarization component

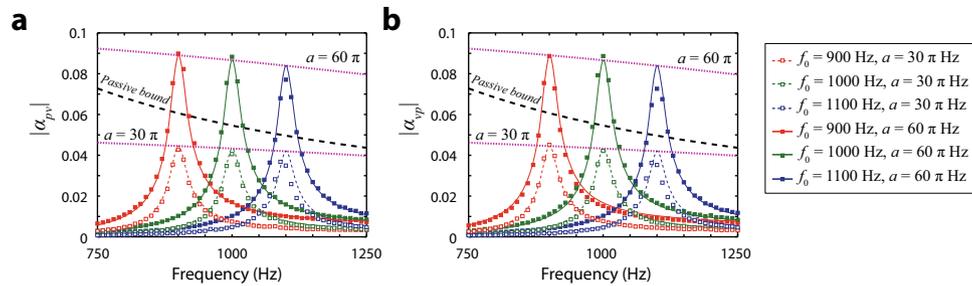


Fig. 4 Willis coupling beyond the passivity bound. Willis coupling beyond the passivity bound achieved by controlling the scaling factors. **a** The magnitude of α_{pv} for the odd-to-even convolution kernel \mathbf{e}_{eo} . The Lorentzian responses at three center frequencies, $f_0 = 900$ Hz (red), 1000 Hz (green) and 1100 Hz (blue), are demonstrated with two different scaling factors, $a = 2\pi \times 15$ (empty symbols) and $2\pi \times 30$ (filled symbols). The analytical results for each scaling factor are also plotted as solid and dashed lines, and the magenta dotted lines denote the theoretical values of the Lorentzian peaks at the resonance frequencies. The black dashed line represents the passivity limit of Willis coupling, i.e., $|\alpha_{pv}| = k_0^{-1}$. **b** Same as **a** except that $|\alpha_{vp}|$ for the even-to-odd convolution kernel \mathbf{e}_{oe} is demonstrated.

is exclusively excited, with the other components suppressed, in excellent agreement with the precise analytical results in Eq. (S23). Since this approach enables simultaneous independent control of the four wave parameters, by balancing the even-to-odd and odd-to-even couplings, we can also easily realize the purely reciprocal and nonreciprocal Willis parameters $\kappa = i(\alpha_{pv} - \alpha_{vp})/2$ and $\chi = (\alpha_{pv} + \alpha_{vp})/2$: from the symmetric convolution kernels $\mathbf{Y} = Y_0(\mathbf{e}_{eo} + \mathbf{e}_{oe})$ for the reciprocal case (Fig. 3e) and the antisymmetric $\mathbf{Y} = Y_0(\mathbf{e}_{eo} - \mathbf{e}_{oe})$ in the nonreciprocal case (Fig. 3f). Our results are not subject to the strict restriction imposed by the geometry of the scatterers in physical metamaterials, in contrast with previous approaches, in which the resonance strengths and bandwidths of individual polarization components are unlikely to be independently configurable. It is further noted that because the virtualized Willis metamaterial can also handle complex polarizabilities, it is possible to achieve controllable gain and loss of the system as well as complex bianisotropy, enabling phenomena such as imaginary reciprocal and nonreciprocal coefficients, which are impossible with conventional bianisotropic media²².

Figure 4 shows the magnitudes of the Willis couplings $|\alpha_{pv}|$ (Fig. 4a) and $|\alpha_{vp}|$ (Fig. 4b) for Lorentzian convolution kernels with two different scaling factors $a = 2\pi \times 15$ (dashed-lines) and $2\pi \times 30$ (solid-lines), respectively, each for below and above the passivity bound at different resonance center frequencies, in agreement with analytical models in Eq. (7). In contrast to the maximum bianisotropy $|\alpha_{vp}|$ (or $|\alpha_{pv}|$) = k_0^{-1} of a passive metamaterial (black dashed line), the newly established maximum bianisotropy for the virtualized Willis metamaterial (magenta dotted line) is modified to $|\alpha_{vp}|$ (or $|\alpha_{pv}|$) = $2k_0^{-1} |\sin(k_0\Delta)Y_0|$ with $|Y_0| = a\gamma^{-1/2}$ at the resonance frequency, revealing the set of parameters for controlling the Willis coupling strength (Supplementary Note S2). The breaking of passive bound for α_{pp} and α_{vv} is also discussed in the Supplementary Note S5. It is worth mentioning that in addition to the control parameter a which represents the power drawn by the active devices, the layout of the scatterers, represented by Δ can also be used to control the strength of the polarizability. However, it is noted that for smaller Δ , polarizabilities become weaker except α_{pp} because the radiation and the detection of dipole mode scale to $\sin(k_0\Delta/2)$ as in Eq. (7). Therefore, the minimum size of Δ and meta-atoms footprint is bound by the targeted magnitude of polarizability, which is set by the dynamic range and sensitivity of active components. In our implementation, a small $\Delta \sim \lambda/7$ was used in the regime of metamaterials.

Inverse design of the broadband Willis response. Recalling that there is no reason for the frequency response of \mathbf{Y} to be restricted to a Lorentzian in our implementation as long as the system

satisfies the causality and stability conditions (see Supplementary Note S4 for the stability analysis), here, we address a metamaterial realization with an arbitrary target response function $F_0(\omega)$ based on the notion of inverse design. To realize $\alpha_{pv}(\omega)$ (or $\alpha_{vp}(\omega)$) = $F_0(\omega)$, we utilize the relation in Eq. (7) and obtain the convolution function $Y_{eo}(\omega) = F_0(\omega)k_0 \sin(k_0\Delta)^{-1/2}$ (or $Y_{oe}(\omega) = -F_0(\omega)k_0 \sin(k_0\Delta)^{-1/2}$) for the target frequency response F_0 . By applying inverse Fourier transformation to $Y_{eo}(\omega)$ (or $Y_{oe}(\omega)$), we can then numerically obtain the required time-domain convolution function $y(t)$. As metamaterials restrict $k_0\Delta$ to be small, the resultant time-domain function of this inverse design process will be similar to the inverse Fourier transform of the original target frequency response $F_0(\omega)$. For example, we consider an intriguing target frequency dispersion with a flat broadband response between ω_1 and ω_2 , specifically,

$$F_0(\omega) = \frac{ae^{i\theta}}{i\pi} \log\left(\frac{\omega + \omega_2}{\omega - \omega_2} \frac{\omega - \omega_1}{\omega + \omega_1}\right), \quad (9)$$

which satisfies the Kramers-Kronig relation, along with its inverse Fourier transform,

$$f_0(t, \theta) = \frac{2a}{\pi t} [\sin(\omega_2 t + \theta) - \sin(\omega_1 t + \theta)] u(t). \quad (10)$$

When ω_2 is set to be much larger than ω_1 , the above function $F_0(\omega)$ with $\theta = 0$ ($\theta = -\pi/2$) provides a flat real (imaginary) spectrum over a broad frequency range while suppressing the imaginary (real) part, while peaks appear in the vicinity of ω_1 and ω_2 . This $F_0(\omega)$ in combination with convolution kernel as used in Fig. 3e, f could then be used to design broadband Willis metamaterials achieving real- or imaginary- κ and χ .

It is worth to mention that, to achieve flat responses, we remove the software-driven derivative in Eq. (4), so the convolution response having the first derivative compensates ω^{-1} dependency of Willis coupling polarizabilities α_{pv} and α_{vp} in Eq. (7). Figure 5 shows the experimental realization of the purely reciprocal Willis parameter κ and the purely nonreciprocal parameter χ , achieving a flat broadband spectrum over $(\omega_1, \omega_2) = (800 \text{ Hz}, 1200 \text{ Hz})$ for $\theta = 0$ (Fig. 5b, c) and $\theta = -\pi/2$ (Fig. 5a, d), with $a = 0.225$.

Here we emphasize that while Fig. 5a, b each correspond to conventional Willis couplings for omega media and moving media, which have real components of κ and χ , respectively, the Willis couplings shown in Fig. 5c, d newly achieve phase-rotated imaginary κ and χ values, providing an additional degree of freedom in terms of energy, i.e., gain and loss in the Willis coupling. While even more general frequency responses can be constructed beyond the Lorentzian resonance and flat dispersion demonstrated here, we note that it is necessary to keep some

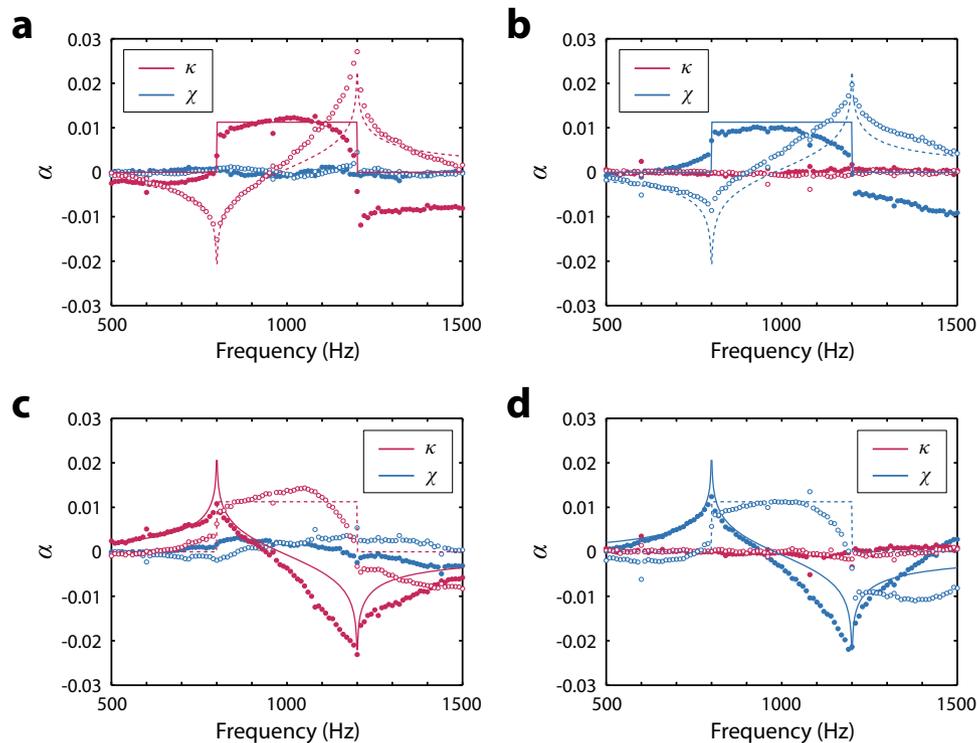


Fig. 5 Broadband frequency dispersion control of Willis couplings. **a, b** Demonstration of broadband frequency dispersion in Willis couplings. Analytically designed target dispersions are calculated with Eq. (9), which provide broadband flat real values. By balancing α_{pv} and α_{vp} , **(a)** broadband reciprocal Willis coupling parameters $\kappa = i(\alpha_{pv} - \alpha_{vp})/2$ and **(b)** broadband nonreciprocal Willis coupling parameters $\chi = (\alpha_{pv} + \alpha_{vp})/2$ are achieved. The polarizabilities κ and χ are depicted in magenta and cyan, respectively, with solid/empty symbols representing the real/imaginary parts of the experimental results. The corresponding analytic models are plotted with solid/dashed lines for the real/imaginary parts. **c, d**, Same as **a** and **b** respectively, except that broadband flat imaginary polarizabilities are demonstrated.

reservations due to the causality restriction. For example, the time-domain convolution function from the inverse Fourier transform of the target frequency response could contain non-causal components, i.e., $y(t) \neq 0$ for $t < 0$ (see Supplementary Note S6 for the flat broadband dispersion of the inverse bulk modulus and mass density) which would require modification of the virtual metamaterial configuration, such as placing the microphones far ahead the speakers.

Discussion

In summary, we demonstrate active Willis meta-atom offering decoupled constitutive parameters beyond the bound of passivity and reciprocity. The conditions for maximum bound and reciprocity in the passivity regime are revisited, and then, the new bound of the constitutive parameters and the reciprocity with the introduction of an active metamaterial are analyzed. By employing a virtualized metamaterial platform that enables the flexible design of scattering properties utilizing constitutive parameter basis representation and software programmed convolution functions, all of the four polarizability parameters are demonstrated independently, which has been impossible for previous electroacoustic metamaterials^{19,28–35}. This capability will open the door to a wider range of applications in non-reciprocal wave manipulation, under the notion of the metasurface. For example, a one-dimensional array of our meta-atoms operated for $t_+ = t_-^{-1} = s$ and $r_+ = r_- = 0$, which is realized with nonreciprocal bianisotropy $\alpha_{vp} = \alpha_{pv} = (s - s^{-1})/2ik_0$ and $\alpha_{pp} = \alpha_{vv} = (s + s^{-1} - 2)/2ik_0$ supports amplification of any incident fields by s times without scattering or phase perturbation. Extension to the two-dimensional application is also possible by placing meta-atom array at the boundary, instead of using

complex potential landscape in the whole interior region³⁹ (see Supplementary Note S7 for the independent control of t_+ and t_-). Exploiting the fully independent excitation of each parameter as well as precise balancing between them, the operations of purely reciprocal and nonreciprocal Willis couplings are realized. We also demonstrate the breaking of the Willis bound in the passivity limit for the first time while isolating the control parameters involved with the newly established Willis bound in the active regime, such as the amplitude, bandwidth, and frequency of the active Lorentzian resonator that feeds in external power for the scattered fields. Extreme Willis coupling could also enable compact design. For example, Li et al.²⁴ showed the zero-index enhanced nonreciprocal metasurfaces in the moving media yet assuming a very large Mach number of ~ 0.1 . It is noted that our platform supports unprecedented nonreciprocity of Mach number of 0.4 (corresponding to experimentally achieved non-reciprocal polarizability of $\chi \sim 0.03$ in Fig. 3f) in addition to the tunability of other acoustic parameters. As well, the number of reciprocal bianisotropic atoms used for extreme impedance matching⁷ could be reduced to 1/6 with the use of our design at $\kappa \sim 0.04$ in Fig. 3e. Finally, we demonstrate the inverse design of flat-amplitude Willis coupling over a broad frequency range, for the reciprocal and nonreciprocal cases as well as the newly revealed case of phase-rotated, nonconserved bianisotropy. Demonstrating full control and top-down tailoring of dispersion, reciprocity, bianisotropy, bulk modulus and mass density within the same platform, our work will enable diverse applications of Willis metamaterials beyond the passivity limit.

Methods

Experimental setup. For the measurement, we used the 4-point measurement method with a National Instruments DAQ device and LabVIEW. Two

microphones, each on the front and back, are placed in an impedance tube and 20 cm and 15 cm away from the specimen. The scattering properties of the meta-atom were tested by means of incident waves coming from the forward and backward directions to identify all four scattering parameters (r_+ , t_+ , r_- , and t_-). In the experimental setup, we flipped the orientation of the meta-atom, while the waves were always incident from the same end of the impedance tube.

Fabrication of the meta-atom. The virtualized meta-atom consists of two MEMS microphones (INMP401) and speakers (SMT-1028-t-2-r) laterally located on each edge of the acrylic frame, which are connected to an external single-board computer (Raspberry Pi 4B+) with amplifiers and analog-to-digital/digital-to-analog converters (see Fig. 2). For digital processing, the input signals sampled by the microphones are digitally processed by the microprocessor and then fed to the speakers in real time with a sampling frequency of $f_s = 7.5$ kHz and a number of samples equal to $N = 400$. The convolution is calculated as $S_j^n[n] = \sum_k \sum_{k=0}^N Y_j[k](M_j[n-k] - M_j[n-k-1])$, where the index $n = t/T_s$ is the discrete time with sampling period $T_s = f_s^{-1}$. The speakers and microphones, which communicate with the microprocessor through the SPI (Serial Peripheral Interface), are mounted in an acrylic frame (width = 3.0 cm, length = 6.5 cm). This transducer module is, in turn, mounted on the acoustic waveguide (width = 3.0 cm, height = 3.0 cm).

Data availability

The data that support the plots in this paper and other findings of this study are available from the corresponding author upon reasonable request.

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Author contributions

J.L. conceived the idea of the virtualization of a Willis metamaterial. C.C. and J.L. established the maximum Willis coupling in a one-dimensional system. C.C. and N.P. established the inverse design of broadband Willis coupling. C.C. and X.W. established the setup of the atom and the control program and performed the measurements. All authors contributed to the data analysis and the writing of the manuscript. J.L. and N.P. managed the project.

Competing interests

The authors declare no competing interests.

Additional information

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