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Highly sensitive low-frequency-detectable acoustic sensor using a piezoresistive cantilever for health monitoring applications

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This study investigates a cantilever-based pressure sensor that can achieve a resolution of approximately 0.2 mPa, over the frequency range of 0.1–250 Hz. A piezoresistive cantilever with ultra-high acoustic compliance is used as the sensing element in the proposed pressure sensor. We achieved a cantilever with a sensitivity of approximately 40 times higher than that of the previous cantilever device by realizing an ultrathin (340 nm thick) structure with large pads and narrow hinges. Based on the measurement results, the proposed pressure sensor can measure acoustic signals with frequencies as low as 0.1 Hz. The proposed pressure sensor can be used to measure low-frequency pressure and sound, which is crucial for various applications, including photoacoustic-based gas/chemical sensing and monitoring of physiological parameters and natural disasters. We demonstrate the measurement of heart sounds with a high SNR of 58 dB. We believe the proposed microphone will be used in various applications, such as wearable health monitoring, monitoring of natural disasters, and realization of high-resolution photoacoustic-based gas sensors. We successfully measured the first (S1) and second (S2) cardiac sounds with frequencies of 7–100 Hz and 20–45 Hz, respectively.

Microphones, which convert sound waves to electrical signals, are indispensable in various applications, such as consumer electronics^{1–5}, automobiles⁶, hearing aid devices^{7,8}, wearable health monitoring^{9–15}, photoacoustic-based gas sensing^{20–25}, and the monitoring of natural disasters, such as volcanic eruptions^{26–28}, debris flow²⁹, and earthquakes³⁰. Two widely used types of microphones exist, traditional electret condenser and microelectromechanical systems (MEMS)-based microphones. Compared with electret condenser microphones, MEMS-based microphones offer better performance at a smaller size¹; thus, they are more suitable in applications where the size of the microphone is a critical design parameter, such as in smartphones, in-ear headsets, wearable devices, and gas-sensing devices.

A major factor that determines the performance of a microphone is the signal-to-noise ratio (SNR), which also indicates the smallest sound pressure that can be detected by the microphone. Significant efforts have been made to improve the SNR of MEMS-based microphones in the recent past, and conventional MEMS-based microphones have an SNR as high as 74 dB³¹, which is equivalent to a minimal detectable sound pressure of approximately 0.32 mPa. Notably, the high SNR of a conventional MEMS-based microphone is obtained only in the audible frequency range (20 Hz–20 kHz), which is the measurement target in most consumer electronics and automotive applications. However, the need for MEMS-based microphones with a high SNR over the low-frequency range (<20 Hz) has recently increased. For example, the frequency of sounds associated with natural disasters is often in the infrasound frequency range (0.01–20 Hz). In health monitoring applications, the desired frequency ranges are 10–250 Hz¹³ for heart sounds and 1–30 Hz for seismocardiography³². In addition, sounds related to respiratory activities can have a frequency component lower than 1 Hz^{16,33}. In non-resonant photoacoustic-based gas sensors^{21,34}, because the photoacoustic pressure increases as the modulation frequency of the incident light decreases³⁵, improving the SNR of sound detection at low frequencies results in a high SNR

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of these sensors. Unfortunately, for frequencies below 20 Hz, the SNR of conventional MEMS-based microphones decreases significantly as the sound frequency decreases^{36–39}.

Thus, this study aims to realize a MEMS-based microphone with a high SNR over the low-frequency range of 0.1–250 Hz. The proposed MEMS-based microphone is based on a nanometer-thick piezoresistive cantilever, as shown in Fig. 1A. When sound pressure is applied to the cantilever, it bends, resulting in a resistance change. Therefore, sound pressure can be detected by measuring the resistance of the cantilever. In previous studies, piezoresistive cantilevers were used to measure the differential pressure and low-frequency sound^{27,40–42}. These cantilevers can maintain a flat frequency response even at frequencies below 1 Hz^{27,41}. However, the SNR of previous piezoresistive cantilevers is still poor owing to its low resolution in response to differential pressure (approximately 20 mPa). In this study, by designing a cantilever with a large pad and narrow, long hinges, high sensitivity of the cantilever in response to sound pressure can be realized, and the minimum detectable pressure as low as 200 μ Pa can be achieved for frequencies as low as 1 Hz. This study investigates the design, fabrication, and performance evaluation of the proposed microphone. Moreover, we demonstrate the measurement of heart sounds using the prototype device.

Design

The design parameters of the cantilever are shown in Fig. 1B. When a differential pressure ΔP is applied to the cantilever, the bending moment caused by the applied pressure $M(x)$ and the area moment of the cantilever $I(x)$ are expressed as follows:

$$M(x) = \begin{cases} \frac{\Delta P w_p}{2} [2l_p(l_h - x) + l_p^2] + \Delta P w_h (l_h - x)^2 & (0 \leq x \leq l_h) \\ \frac{\Delta P w_p (l_h + l_p - x)^2}{2} & (l_h \leq x \leq l_h + l_p) \end{cases} \quad (1)$$

$$I(x) = \begin{cases} \frac{w_h t_c^3}{6} & (0 \leq x \leq l_h) \\ \frac{w_p t_c^3}{12} & (l_h \leq x \leq l_h + l_p) \end{cases} \quad (2)$$

where t_c , w_h , w_p , l_h , and l_p are the thickness of the cantilever, width of the cantilever hinge, width of the cantilever's pad, length of the cantilever hinge, and length of the cantilever's pad, respectively. The cantilever deformation profile for a minor deflection $y(x)$ is expressed as follows:

$$\frac{d^2 y}{dx^2} = \frac{M(x)}{E \cdot I(x)} \quad (3)$$

where E is Young's modulus of the cantilever material. The boundary conditions when the x -axis originates from the fixed end of the cantilever, as shown in Fig. 1, are expressed as follows:

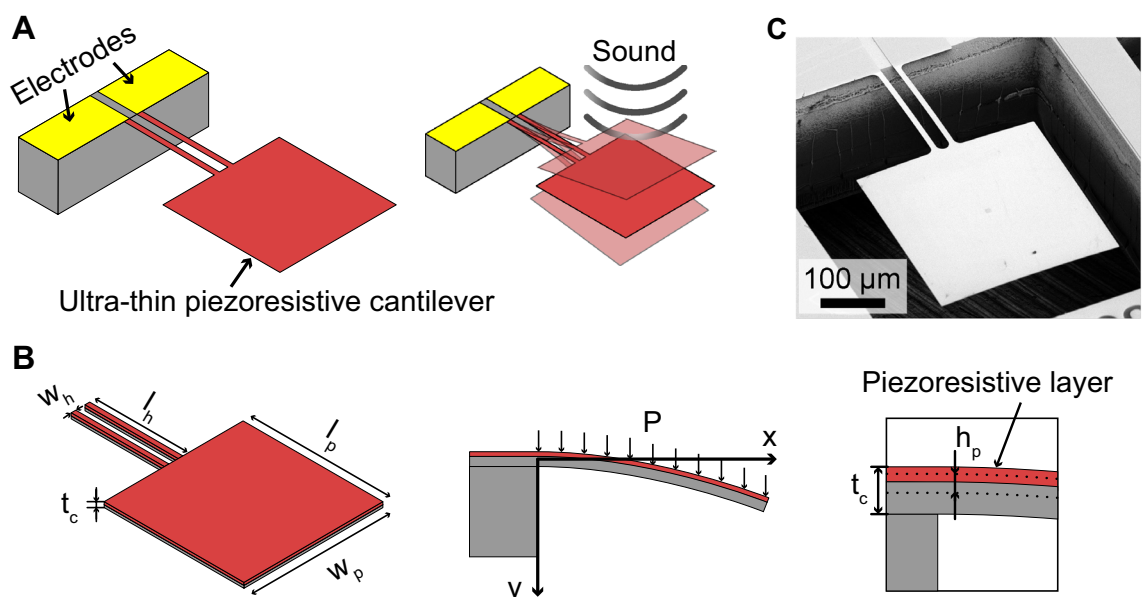


Figure 1. (A) Conceptual illustration of the proposed microphone using a piezoresistive cantilever. (B) Definition of the design parameters of the cantilever. (C) SEM image of the fabricated cantilever.

$$\frac{dy(0)}{dx} = 0 \tag{4}$$

$$y(0) = 0 \tag{5}$$

The deflection $y(x)$ and the spring constant k can be calculated from these equations. From Eqs. (1), (2), (3), and (5), the deflection of the cantilever $y(x)$ when the constant pressure ΔP is applied to the surface is described as follows:

$$\frac{Et_c^3 w_h(x)}{6} \frac{d^2 y}{dx^2} = \Delta P w_h (l_h - x)^2 + \frac{\Delta P w_p (-l_p^2 + 2l_p(l_h + l_p - x))}{2} \quad (0 \leq x \leq l_h) \tag{6}$$

$$\frac{Et_c^3 w_p(x)}{12} \frac{d^2 y}{dx^2} = \frac{\Delta P w_p (l_h + l_p - x)^2}{2} \quad (l_h \leq x \leq l_p + l_h) \tag{7}$$

Therefore, from Eqs. (6) and (7), the deflection $y(x)$ can be mathematically derived as follows:

$$y(x) = \begin{cases} \frac{\Delta P x^2 \cdot (6l_h^2 w_h - 4l_h w_h x - 2l_p w_p x + 3l_p w_p (2l_h + l_p) + w_h x^2)}{2Et_c^3 w_h} & (0 \leq x \leq l_h) \\ \frac{\Delta P (2l_p^2 l_p w_h \cdot (2l_h + 3l_p) - l_p^2 l_p w_p \cdot (2l_h + 3l_p) - 12l_h l_p w_h x (l_h + l_p) + 6l_h l_p w_p x (l_h + l_p) + w_h x^4 - 4w_h x^3 (l_h + l_p) + 6w_h x^2 (l_h + l_p)^2)}{2Et_c^3 w_h} & (l_h \leq x \leq l_h + l_p) \end{cases} \tag{8}$$

The maximum deflection of the cantilever at $x = l_h + l_p$ is

$$d = y(x = l_h + l_p) = \frac{\Delta P (l_h l_p w_p \cdot (4l_h^2 + 9l_h l_p + 6l_p^2) + w_h \cdot (3l_h^4 + 4l_h^3 l_p + 3l_p^4))}{2Et_c^3 w_h} \tag{9}$$

Therefore, using Eq. (9), the spring constant when constant pressure is uniformly applied to the cantilever is expressed as follows:

$$k = \frac{\Delta P (2w_h l_h + w_p l_p)}{d} = \frac{2Et_c^3 w_h \cdot (2l_h w_h + l_p w_p)}{l_h l_p w_p \cdot (4l_h^2 + 9l_h l_p + 6l_p^2) + w_h \cdot (3l_h^4 + 4l_h^3 l_p + 3l_p^4)} \tag{10}$$

Conversely, from⁴³, the spring constant k when a concentrated load is applied at the tip of the cantilever is expressed as:

$$k = \frac{E w_p w_h t_c^3}{4w_h l_p^3 + 2t_c (3l_h^2 l_p + 3l_h l_p^2 + l_h^3)} \tag{11}$$

As a dynamic property, the first resonance frequency of the cantilever is expressed as:

$$f_r = \frac{1}{2\pi} \sqrt{\frac{k}{m_{eff}}} \tag{12}$$

where k and m_{eff} are the spring constant and effective mass of the cantilever, respectively. In this structure, m_{eff} is expressed as $0.24m$, where m is the cantilever mass. Considering the damping ratio in air, the resonant frequency in air, f_d is expressed as:

$$f_d = \sqrt{1 - \xi^2} f_r \tag{13}$$

The design parameters of the cantilever used in this study are listed in Table 1. As observed in equation (9), the wider square pad (w_p and l_p) should be designed for larger displacement per the same applied pressure (ΔP). However, these increases also result in the resonant frequency reduction, as observed in Eqs. (10), (12), and (13). In our target of heart-sound monitoring, the maximum frequency is less than 300 Hz⁴⁴. Therefore, We designed w_p and l_p to be sensitive to signals up to 300 Hz and three times larger than in previous studies reported by Takahashi et al.⁴⁰ to obtain larger signals.

As an electrical property of the piezoresistive layer, the fractional resistance change $\Delta R/R$ of the cantilever can be calculated as⁴⁰

$$\frac{\Delta R}{R} = \frac{\alpha h_p}{l_h} \int_0^{l_h} \frac{M(x)}{I(x)} dx \tag{14}$$

where α is a constant indicating the relationship between the fractional resistance change and stress, and h_p is the distance between the piezoresistive layer and the neutral axis (x -axis, as shown in Fig. 1B). Inserting Eqs. (1) and (2) into Eq. (14), we obtain

$$\frac{\Delta R}{R} = \Delta P h_p \frac{2l_h^2 w_h + 3l_p w_p (l_h + l_p)}{t_c^3 w_h} \quad (15)$$

As observed in Eq. (15), for the same cantilever thickness t_c and piezoresistive layer, the sensitivity of the cantilever $(\Delta R/R)/\Delta P$ can be improved by increasing the length l_h of the cantilever hinges, the width w_p and length l_p of the cantilever pad, and decreasing the width w_h of the cantilever hinges. In other words, high sensitivity can be achieved by designing a cantilever with long narrow hinges and a wide pad.

Using Eq. (15) and assuming that α and h_p are the same for both designs, the sensitivity of the designed cantilever is expected to be approximately 55 times higher than that of the previous cantilever design⁴⁰. The finite element method (FEM) simulation results for the cantilever using ANSYS are shown in Fig. 2A. Under vacuum conditions, the resonant frequency is 498 Hz. ξ was swept from 0.1 to 0.6 to investigate the damping effect. As the value of ξ increases, the resonant peak flattens, as shown in Fig. 2B.

Results

Measurement of resonant frequency. The first resonant frequency of the cantilever was evaluated using the setup shown in Fig. 3A. The fabricated sensor chip was attached to a vibrator (type 4810, Brüel & Kjær, Nærum, Denmark), which applied vibration in the frequency range of 10 Hz to 600 Hz to the sensor chip. A laser Doppler vibrometer (VFX-I-130, Polytec, Baden-Württemberg, Germany) was used to measure the vibrations of the cantilever surface and the base where the handle and device layer of the silicon-on-insulator wafer were not etched. As shown in Fig. 3A, the base vibrator had a resonant frequency of 100 Hz. Therefore, the cantilever's

w_h	l_h	w_p	l_p	t_c
10 μm	20 μm	300 μm	300 μm	0.34 μm

Table 1. Design parameters of the cantilever.

signal also had peaks at 100 Hz. To eliminate the effect of the base vibrator's resonance, we measured the ratio of the cantilever and the base vibrator outputs. The measurement results are shown in Fig. 3B, which describes the ratio of the cantilever's vibration amplitude to that of the chip base. From the result, the first resonant frequency of the cantilever was determined to be approximately 275 Hz. Therefore, the cantilever is suitable for measuring sound waves with frequencies as high as 275 Hz. Then, we evaluated the cantilever's vibration when the sound waves were applied. For the measurement, the fabricated cantilever and the reference microphone were placed at the same distance of 1 cm from a loudspeaker (KFC-XS174S, Kenwood Corp., Tokyo, Japan). The vibration of the cantilever was measured using the same vibrator. Figure 3C shows the raw data of the cantilever's velocity and the reference microphone's output. Figure 3D shows the frequency responses of the cantilever's vibration amplitude when the sound waves were applied using the loudspeaker. The measurement results corresponded to the simulation results under damped conditions, as shown in Fig. 2B. We supposed the sensor was utilized in the atmosphere, and the air-damping factor was constant. Due to the damping, we obtained flat frequency responses over a frequency range of 1–250 Hz without the sharp resonant peak.

Performance evaluation. We evaluated the sensitivity and SNR of the fabricated cantilever in response to sound waves using the experimental setup shown in Fig. 4A. Photographs of the experimental setup are shown in Fig. 4B. For the measurement, the fabricated cantilever and a reference microphone (Type 4955, Brüel & Kjær, Nærum, Denmark) were placed at the same distance of 1 cm from a loudspeaker (KFC-XS174S, Kenwood Corp.,

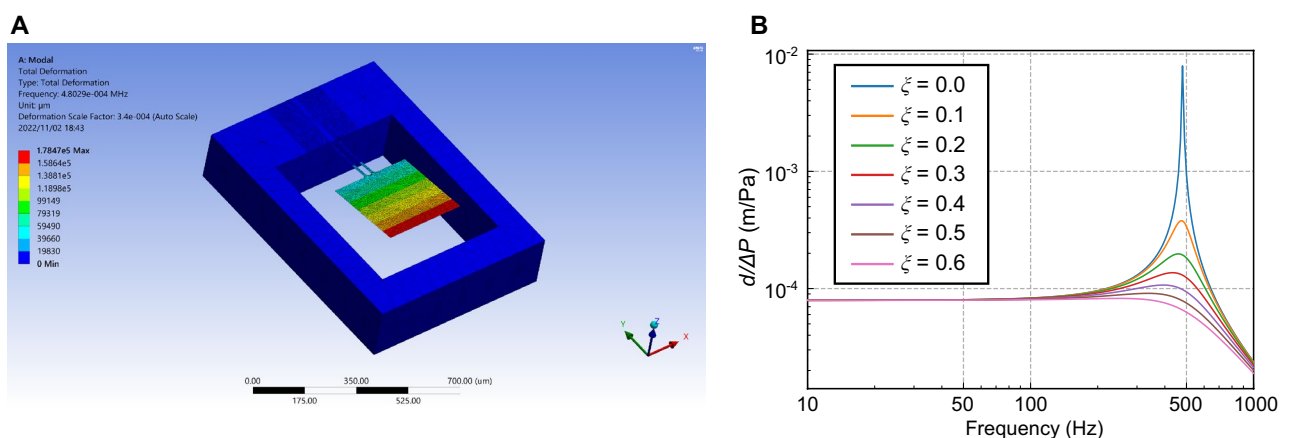


Figure 2. (A) FEM simulation of the proposed cantilever. (B) FEM simulation results sweeping the damping factor.

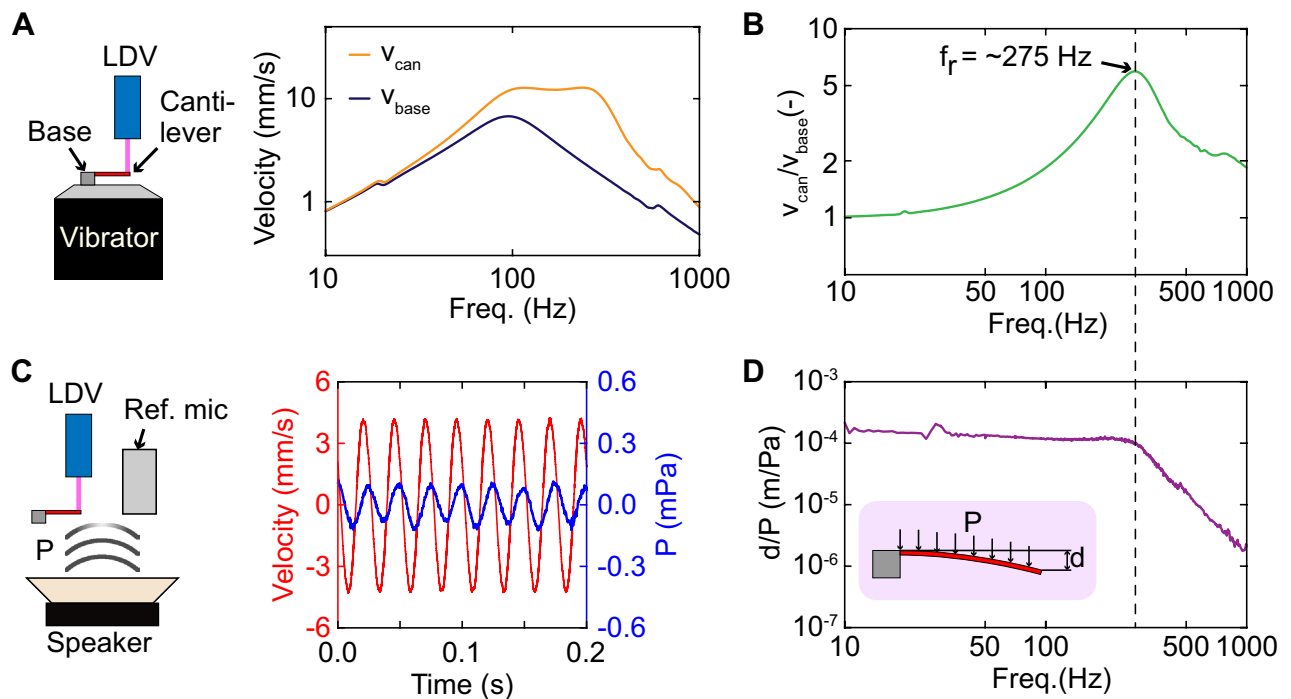


Figure 3. (A) Setup to measure the resonant frequency of the cantilever. (B) The frequency response characteristic of the cantilever in response to the vibration applied to the chip. V_{can} and V_{base} represent the vibration amplitudes of the cantilever and the chip base, respectively. (C) Raw data of the cantilever's vibration and reference microphone output. (D) Frequency spectrum data of the cantilever's vibration amplitude when the sound waves are applied.

Tokyo, Japan). The evaluation setup was placed inside an anechoic chamber to reduce ambient acoustic noise. The loudspeaker was driven by applying an AC voltage from a frequency response analyzer (FRA51602, NF, Yokohama, Japan). The resistance change of the cantilever was measured using a lab-made amplifier composed of a Wheatstone bridge circuit. The resistance change of the cantilever was calculated from the output of the voltage change of the amplifier output, as follows:

$$\frac{\Delta R}{R} = \frac{4 \Delta V}{G V_0} \quad (16)$$

where ΔV is the voltage change in the amplifier, $G = 5000$ is the amplifier gain, and $V_0 = 1 \text{ V}$ is the voltage applied to the Wheatstone bridge circuit. The outputs of the amplifier and reference microphone were simultaneously measured using a frequency response analyzer. The sound pressure applied to the cantilever can be calculated from the output of the microphone, and the sensitivity of the cantilever, defined as the ratio between the resistance change and applied sound pressure, can be calculated as $(\Delta R/R)/P$.

The measured sensitivity of the fabricated cantilever and its noise spectrum density (NSD) over a frequency range of 1–1000 Hz is shown in Fig. 4C. We performed the measurement five times. The red line represents the average of five measurements. The NSD was calculated as $(\Delta R/R)$ divided by the square root of the bandwidth when the speaker was switched off. In this experiment, we performed the experiment without using the laser irradiation to the cantilever inside an anechoic chamber. Besides, the power consumption in the piezoresistance in the cantilever is small enough to be ignored, and the room temperature was constant. Therefore, we consider that the dominant noise factor is caused by the amplifier. The NSD caused by the amplifier was $50 \mu\text{V} \cdot \text{Hz}^{-1/2}$ at 10 Hz and is inversely proportional to the frequency of the input signal. This equivalent value converted to piezoresistive change $(\Delta R/R)$ is $4.0 \times 10^{-5} \text{ Hz}^{-1/2}$, which meets the measured NSD, as shown in Fig. 4C. Based on the sensitivity measurement result, the cantilever exhibits a flat frequency response in the frequency range of 1 to over 250 Hz. When the frequency increased in the range of over 250 Hz, the sensitivity of the cantilever started decreasing. The measured sensitivity was unstable at low frequencies ($< 5 \text{ Hz}$). This fluctuation in the sensitivity of the cantilever in the low-frequency region resulted from the low output signal of the reference microphone and the limitation of the loudspeaker in this frequency range. Over the frequency range of 5 to 250 Hz, the average sensitivity of the cantilever was $12.4 \times 10^{-3} \text{ Pa}^{-1}$, approximately 40 times higher than the previous cantilever with a similar thickness⁴⁰. This improvement in sensitivity is also of the same order as the result of the theoretical analysis described in the previous section. In contrast, the NSD of the cantilever resistance change calculated from the amplifier circuit output was generally less than $2.0 \times 10^{-6} \text{ Hz}^{-1/2}$ over the measured frequency ranges. From the measured sensitivity and NSD, SNR is calculated as:

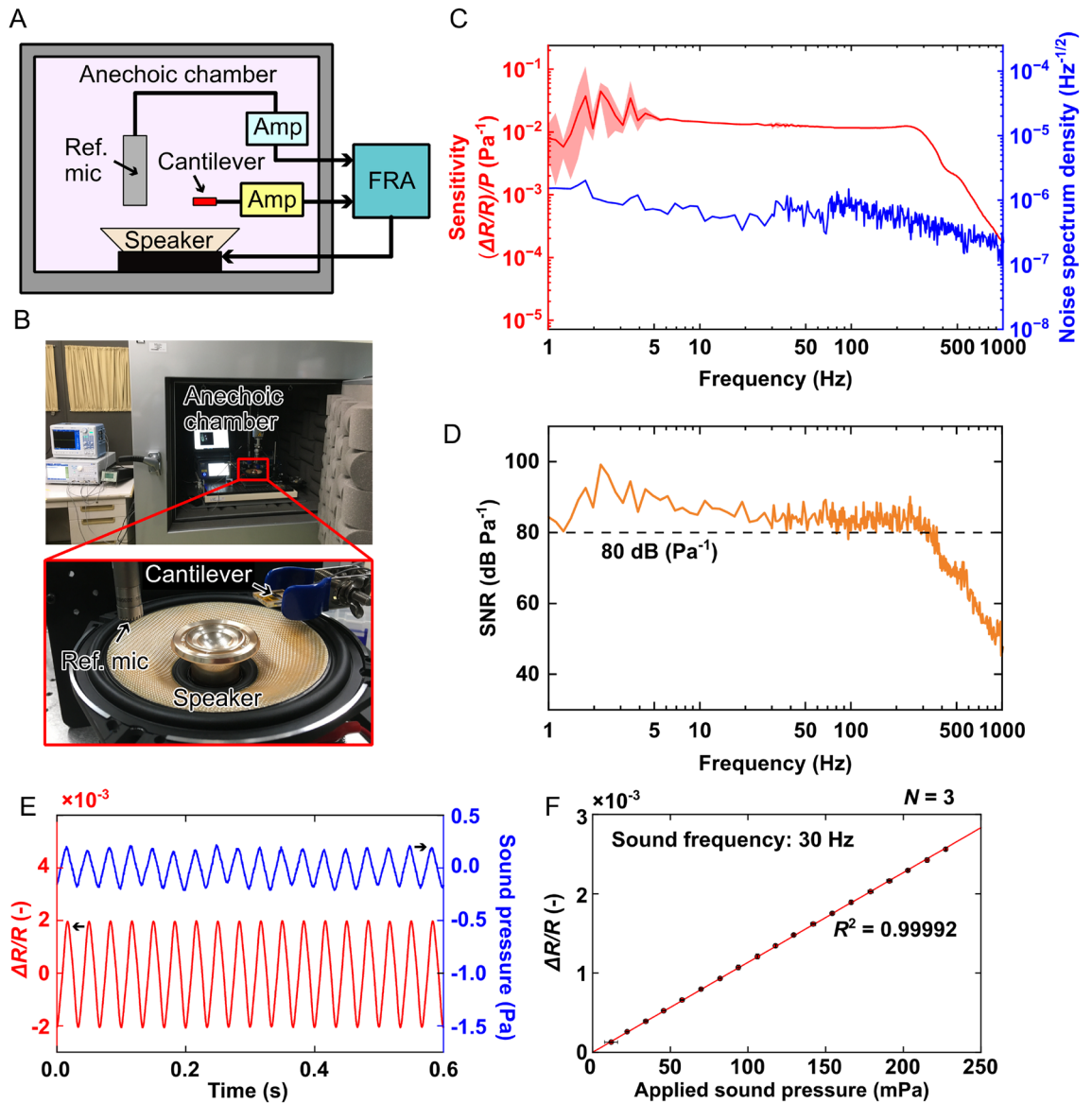


Figure 4. (A) Conceptual design of the experimental setup to evaluate the performance of the cantilever (B) Photographs of the experimental setup. (C) Measurement results of the sensitivity and noise spectrum of the cantilever output. (D) The resolution and SNR derived from the results in (C). (E) Raw data of the cantilever and reference microphone output. (F) Relationship between the resistance change of the cantilever and the applied sound pressure.

$$\begin{aligned}
 \text{SNR} &= \frac{\text{Sensitivity} \cdot \sqrt{f_{\text{BW}}}}{\text{NSD} \cdot \sqrt{f_{\text{BW}}}} \\
 &= \frac{\Delta R/R/P}{(\Delta R/R)_{\text{noise}}}
 \end{aligned}
 \tag{17}$$

where f_{BW} is the measurement bandwidth. Figure 4D shows the derived SNR of the cantilever. The SNR is approximately 80 dB Pa⁻¹ over a frequency range of 1–250 Hz.

Moreover, we evaluated the linearity of the cantilever output by changing the applied sound pressure and measuring the change in cantilever resistance. During the measurement, the frequency of the applied sound was fixed at 30 Hz, and the voltage applied to the speaker was varied in the range of 0.5–10 V. An example of the measured data is shown in Fig. 4E. The sound pressure was calculated directly from the measured signal from the reference microphone. The relationship between the change in resistance of the cantilever and the applied sound pressure is shown in Fig. 4F. The applied pressure ranged from 0 to 250 mPa. The resistance change of the cantilever increased linearly with an increasing sound pressure with a coefficient of $11.3 \times 10^{-3} \text{ Pa}^{-1}$, which is in good agreement with the sensitivity shown in Fig. 4C.

To estimate the minimum detectable pressure, the lower limit of the detectable piezoresistive change is defined as 3 dB higher than the noise value. In Fig. 5A, the 3 dB larger piezoresistive change ($(\Delta R/R)_{\text{Noise}+3\text{dB}}$)

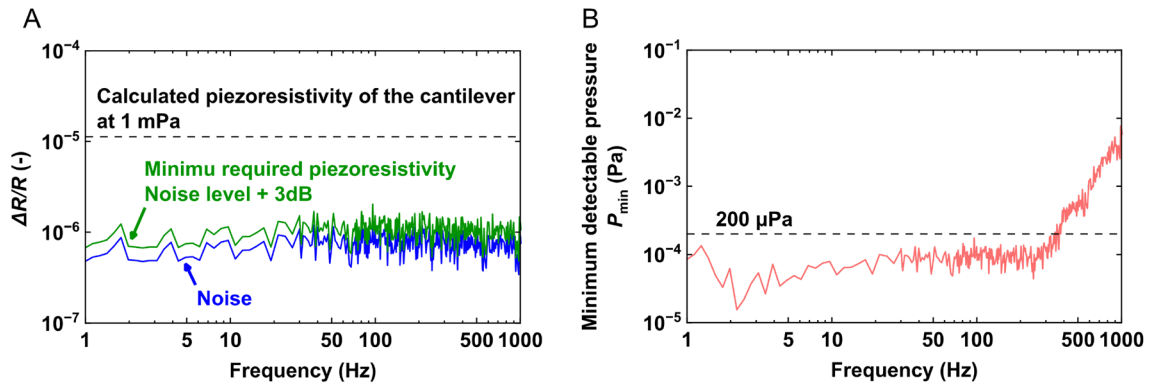


Figure 5. (A) The piezoresistivity of the noise and 3 dB higher than the noise level. The dashed line represents the calculated piezoresistivity at 1 mPa pressure (B) Minimum detectable pressure (P_{\min}) calculated from the sensitivity.

is represented by the green line, and the blue line represents the noise level. The dashed line represents the piezoresistivity calculated from result shown in Fig. 4F, when 1-mPa pressure is applied. As shown in Fig. 5A, the resistive change of the proposed sensor at 1 mPa is much higher to the minimum required value. Figure 5B shows the minimum detectable pressure (P_{\min}) calculated as:

$$P_{\min} = \frac{(\Delta R/R)_{\text{Noise}+3\text{dB}}}{\text{Resolution}} = \frac{(\Delta R/R)_{\text{Noise}+3\text{dB}}}{(\Delta R/R/P)_{\text{Sens}}} \quad (18)$$

where $(\Delta R/R/P)_{\text{Sens}}$ is the sensitivity of the proposed cantilever, as shown in Fig. 4C. Based on the results, P_{\min} of the proposed cantilever was approximately 200 μPa over a frequency range of 1–250 Hz, which is the highest among MEMS-based microphones.

From the results shown in Fig. 4, the sensitivity of the cantilever's output was considerably better than that of the reference microphone's output, particularly at low frequencies. When the frequency increased over 250 Hz, the sensitivity of the cantilever started decreasing. The measured sensitivity was unstable at low frequencies (< 5 Hz). For frequencies below 5 Hz, the signals of the reference microphone could not be distinguished from its noise, whereas the signal of the cantilever could be observed in the raw data and frequency spectra. Because the sensitivity of the cantilever shown in Fig. 4C was derived by dividing the output of the cantilever by that of the reference microphone, the low SNR of the reference microphone in the frequency range of 1–5 Hz caused the fluctuation in the calculated sensitivity of the cantilever over this frequency range.

We performed another measurement using the differential air pressure to investigate the response of the cantilever at a frequency lower than 5 Hz, as shown in Fig. 6A. The cantilever device was connected to an air chamber where the pressure was controlled using a syringe. In this measurement, the gain of amplifier G was set to 1000, and the voltage change of amplifier V_0 was set to 1 V. The differential pressure was controlled by changing the actuator position connected to the syringe. In the previous study⁴¹, the measurement system investigated the device characteristics at a low frequency of less than 10 Hz. We connected a reference pressure sensor to validate the application of differential pressure. The measurement results of the proposed acoustic sensor and the referential pressure sensor, as well as their frequency spectra, obtained using fast Fourier transform are shown in Fig. 6B. Based on the results, the proposed cantilever sensor is sensitive even in the low-frequency range, and the results are consistent with the results of the sound wave experiment, as shown in Fig. 4.

Heart sound measurement. We measured heart sounds using the fabricated cantilever to demonstrate the application of the proposed microphone in health monitoring. The fabricated cantilever chip was glued to a 3D-printed jig attached to the chest of a subject (male, 37 years old), as shown in Fig. 7A. The thickness of the jig allowed the cantilever to be located approximately 3 mm above the subject's chest. The raw data of a measurement performed for 50 s is shown in Fig. 7B. The cantilever can measure the heart sound of the subject, and the peak-to-peak amplitude of the heart sound signal was approximately 8×10^{-4} (-). As shown in Fig. 7C, the noise level of the amplifier is equivalent to a fractional resistance change of 10^{-6} (-); thus, the recorded heart sound signal has an SNR of 58 dB, which is 17 dB better than the measurement result obtained using a highly sensitive mechanoacoustic sensor based on the nanofiber previously reported¹³.

A zoomed-in view of the recorded heart sound in the duration of 15 to 17 s and its wavelet scalogram are shown in Fig. 7C,D shows the raw signal in the same duration. As shown in Fig. 7C,D, the signal has two dominant events. These events are well-known heart sounds, called as the first (S1) and second (S2) cardiac sounds^{44,45}. S1 is generated by the closure of the mitral and tricuspid valves, and the closure of the aortic and pulmonary valves generates S2. Figure 7E shows the zoomed-in view of a single cycle of the heart sound. As shown in Fig. 7e, the major components in S1 and S2 were observed. In S1, three peaks were observed. The first two sharp peaks (M1 and T1) were caused by the closure of the mitral and tricuspid valves, respectively⁴⁴. In this measurement,

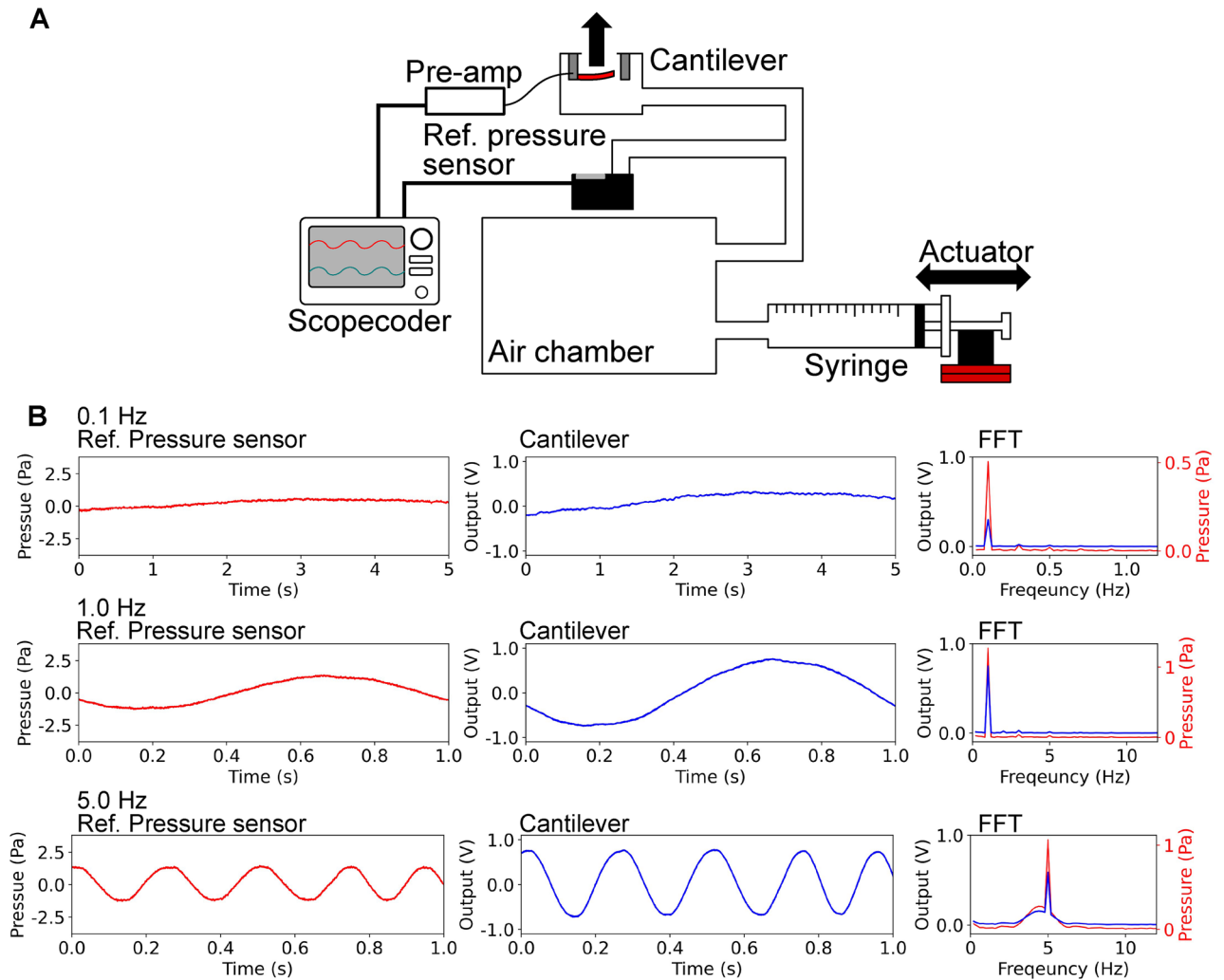


Figure 6. (A) Conceptual schematic of the experimental setup to evaluate the performance of the cantilever at low frequencies. (B) Timescale and frequency spectrum data of the measurement results of the reference pressure sensor and the proposed acoustic sensor.

T1 was heard 28 msec later than M1. The delay is known to be smaller than 35 msec, and the value is normal. The final peak (E) was the ejection sound associated with the opening of the aortic valve. In S2, two peaks were observed. The first sound (A2) was generated by the closure of the aortic valve, and the pulmonary valve generated the second sound (P2). The delay between the A2 and P2 was 63 msec.

Moreover, from the wavelet scalogram, the dominant frequency ranges of S1 and S2 were 7 to 100 and 20 to 45 Hz, respectively. In addition, the high SNR of the recorded signal facilitated the calculation of the timing of the peak of S1 for each heartbeat. From this calculation, the heart rate of the subject can be obtained for each heartbeat, as shown in Fig. 7F. Based on the aforementioned measurement results, the proposed cantilever is an excellent candidate for wearable heart sound monitoring devices. The proposed cantilever can measure heart sounds when placed several millimeters above the skin. In other words, the cantilever must not be attached to the skin to measure the heart sound, as required by many mechanoacoustic sensors. This is advantageous for wearable health monitoring devices because the contact area between the device and the skin is significantly reduced, causing discomfort to the user.

Methods

Fabrication. The fabrication process of the sensor is shown in Fig. 8. A silicon-on-insulator wafer (thickness: 0.34/0.4/250 μm) was used to fabricate the sensor. First, the piezoresistive layer was formed by doping the device layer with arsenic ions via ion implantation followed by annealing (700°C for 10 min, and by 900°C for 30 min). Next, the oxide layer formed on the Si layer during annealing was removed by dipping the wafer in a hydrogen fluoride (HF) solution. Chromium and gold layers with thicknesses of 5 and 50 nm were deposited on the piezoresistive layer, respectively. The metal and Si layers were then etched by wet etching and deep reactive ion etching, respectively, to form the cantilever. Next, the metal layers on top of the cantilever were removed by wet etching. Finally, the cantilever was released by etching the handle Si layer and removing the box layer using

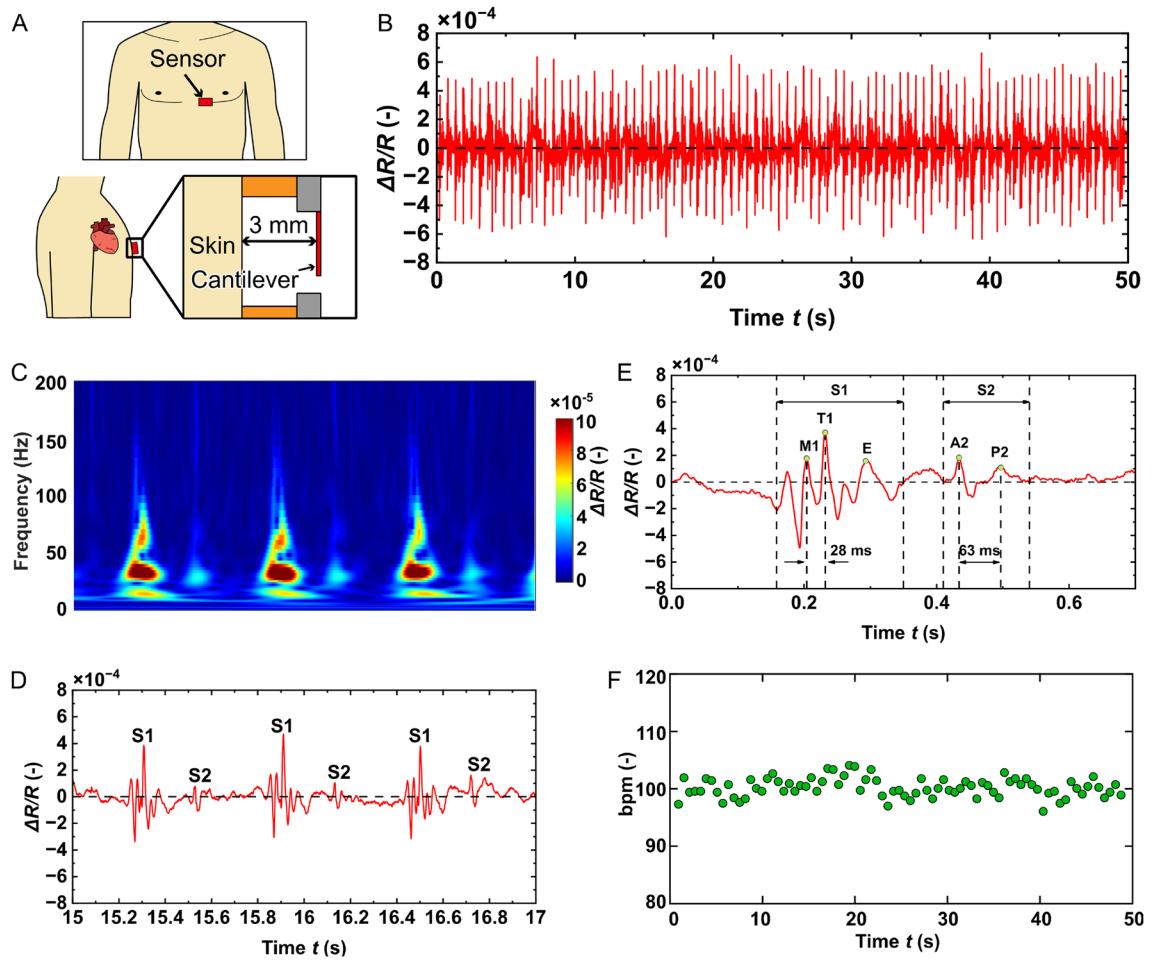


Figure 7. (A) Conceptual schematic of the proposed sensor for measuring heart sound. (B) Raw data of the pulse sound measurement. (C) Wavelet scalogram of the recorded heart sound in the duration of 15–17 s. (D) A zoomed-in view of the recorded heart sound in the duration of 15–17 s. (E) A zoomed-in view of a single cycle of the heart sound. (F) Heart rate of the subject for each individual heartbeat.

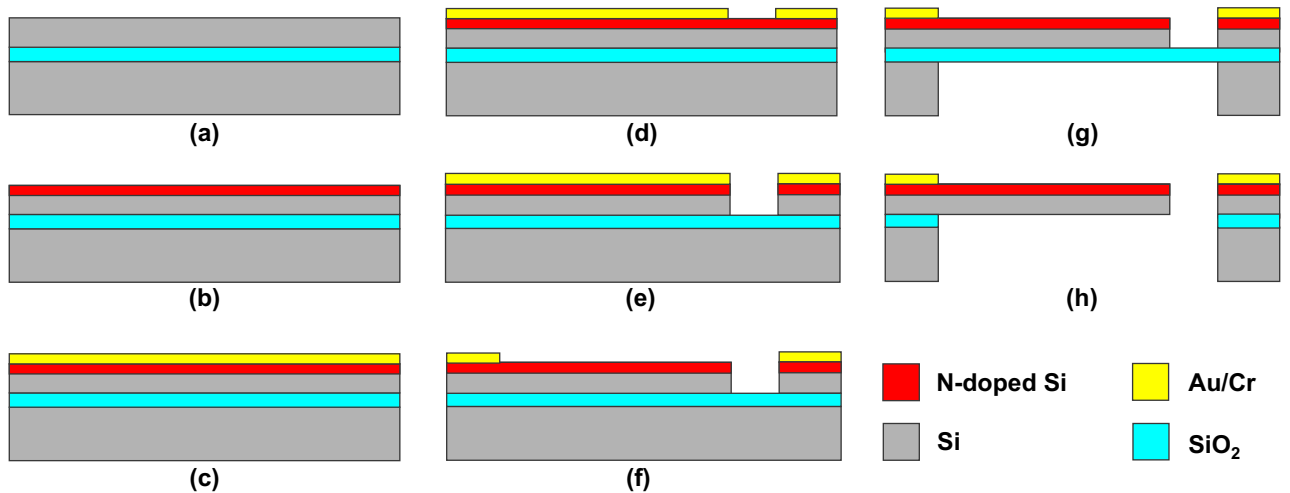


Figure 8. Fabrication process of the proposed cantilever-type acoustic sensor.

HF vapor. Moreover, additional cleaning with O₂ plasma was applied to remove the residual photoresist and oil attached to the cantilever after removing the box layer. The SEM image of the fabricated cantilever is shown in Fig. 1C. The initial resistance of the cantilever was approximately 5.4 kΩ.

Heart sound measurement. The jig attached to the chest of a subject was fabricated using an inkjet 3D printer (AGILISTA, Keyence, Japan) with a minimum pattern width of 15 μm. The authors declare that the jig does not have skin-irritating properties. The experimental protocols in this paper were approved by the IRB of Ethics in Ergonomic Experiments (2019-988-B) in the National Institute of Advanced Industrial Science and Technology (AIST), including any relevant details. Moreover, the authors declare that all experiments were carried out following the relevant guidance and regulations and confirm that informed consent was obtained from all participants.

Conclusions

This study investigated the design of a MEMS-based microphone that can achieve a resolution of approximately 0.2 mPa over the frequency range of 0.1 to 250 Hz, which is the highest value reported to date for MEMS-based acoustic sensors. The high performance of the proposed microphone was enabled by the piezoresistive cantilever, which was 340 nm thick and possessed a 300 μm-wide pad and two 10 μm-wide, 200 μm-long hinges. This extremely flexible structure allowed the cantilever to obtain a sensitivity of over 10⁻² Pa⁻¹, which was 40 times higher than that of the previous cantilever design with a similar thickness. The superior SNR of the cantilever over the low-frequency sound compared with the reference commercial microphone was demonstrated through measurement results. Using the fabricated cantilever, we demonstrated the measurement of heart sounds with an SNR as high as 58 dB. We believe that the proposed microphone will be useful in various applications, such as wearable health monitoring, monitoring of natural disasters, and realization of high-resolution photoacoustic-based gas sensors.

Data availability

The datasets generated during and/or analysed during the current study are available from the corresponding author on reasonable request.

Received: 22 December 2022; Accepted: 14 April 2023

Published online: 20 April 2023

References

1. Tilli, M. *et al.* *Handbook of Silicon Based MEMS Materials and Technologies* (Elsevier, 2020).
2. Zawawi, S. A., Hamzah, A. A., Majlis, B. Y. & Mohd-Yasin, F. A review of mems capacitive microphones. *Micromachines* **11**, 484 (2020).
3. Zinslering, B. Silicon-based mems microphone for automotive applications. *Micronano News* 8–11 (2007).
4. Bogue, R. Recent developments in mems sensors: A review of applications, markets and technologies. *Sens. Rev.* 300–304 (2013).
5. Ali, W. R. & Prasad, M. Piezoelectric mems based acoustic sensors: A review. *Sens. Actuators, A Phys.* **301**, 111756 (2020).
6. Yoo, I. *et al.* Development of directional mems microphone single module for high directivity and snr. *IEEE Sens. J.* **22**, 6329–6335 (2021).
7. Mallik, S., Chowdhury, D. & Chhrtopadhyay, M. Development and performance analysis of a low-cost mems microphone-based hearing aid with three different audio amplifiers. *Innov. Syst. Softw. Eng.* **15**, 17–25 (2019).
8. Zargarpour, N. & Zarifi, M. H. A piezoelectric micro-electromechanical microphone for implantable hearing aid applications. *Microsyst. Technol.* **21**, 893–902 (2015).
9. Cui, J. *et al.* Design and optimization of mems heart sound sensor based on bionic structure. *Sens. Actuators, A Phys.* **333**, 113188 (2022).
10. Zhang, G., Liu, M., Guo, N. & Zhang, W. Design of the mems piezoresistive electronic heart sound sensor. *Sensors* **16**, 1728 (2016).
11. Lee, S. H., Kim, Y.-S. & Yeo, W.-H. Advances in microsensors and wearable bioelectronics for digital stethoscopes in health monitoring and disease diagnosis. *Adv. Healthcare Mater.* **10**, 2101400 (2021).
12. Wang, H., Chen, J., Hu, Y., Jiang, Z. & Samjin, C. Heart sound measurement and analysis system with digital stethoscope. In *2009 2nd International Conference on Biomedical Engineering and Informatics*, 1–5 (2009).
13. Nayeem, M. O. G. *et al.* All-nanofiber-based, ultrasensitive, gas-permeable mechanoacoustic sensors for continuous long-term heart monitoring. *Proc. Natl. Acad. Sci.* **117**, 7063–7070 (2020).
14. Qu, M. *et al.* Heart sound monitoring based on a piezoelectric mems acoustic sensor. In *2021 IEEE 34th International Conference on Micro Electro Mechanical Systems (MEMS)*, 59–63 (2021).
15. Nguyen, T.-V. & Ichiki, M. Simultaneous measurement of pulse wave and respiration using a single tube-shaped mems-based pressure sensor. In *2020 IEEE 33rd International Conference on Micro Electro Mechanical Systems (MEMS)*, 84–87 (2020).
16. Nguyen, T.-V., Okada, H., Takei, Y., Takei, A. & Ichiki, M. A band-aid type sensor for wearable physiological monitoring. In *2021 21st International Conference on Solid-State Sensors, Actuators and Microsystems (Transducers)*, 1432–1435 (2021).
17. Nguyen, T.-V. & Ichiki, M. MemS-based sensor for simultaneous measurement of pulse wave and respiration rate. *Sensors* **19**, 4942 (2019).
18. Nguyen, T.-V. & Ichiki, M. Mask-type sensor for pulse wave and respiration measurements and eye blink detection. *Sensors* **21**, 4895 (2021).
19. Nguyen, T.-V. *et al.* MemS-based pulse wave sensor utilizing a piezoresistive cantilever. *Sensors* **20**, 1052 (2020).
20. Chen, K. *et al.* Highly sensitive photoacoustic microcavity gas sensor for leak detection. *Sensors* **20**, 1164 (2020).
21. Chen, K. *et al.* Highly sensitive photoacoustic gas sensor based on multiple reflections on the cell wall. *Sens. Actuators A* **290**, 119–124 (2019).
22. Hopper, R., Popa, D., Udrea, F., Ali, S. Z. & Stanley-Marbell, P. Miniaturized thermal acoustic gas sensor based on a cmos micro-hotplate and mems microphone. *Sci. Rep.* **12**, 1–6 (2022).
23. Gong, Z. *et al.* High-sensitivity Fabry–Perot interferometric acoustic sensor for low-frequency acoustic pressure detections. *J. Lightwave Technol.* **35**, 5276–5279 (2017).
24. Gong, Z. *et al.* Parylene-c diaphragm-based low-frequency photoacoustic sensor for space-limited trace gas detection. *Opt. Lasers Eng.* **134**, 106288 (2020).

25. Fu, X. *et al.* Micromachined extrinsic Fabry–Pérot cavity for low-frequency acoustic wave sensing. *Opt. Express* **27**, 24300–24310 (2019).
26. Johnson, J. B. & Ripepe, M. Volcano infrasound: A review. *J. Volcanol. Geoth. Res.* **206**, 61–69 (2011).
27. Shimatani, J., Takahashi, H., Ichihara, M., Takahata, T. & Shimoyama, I. Monitoring volcanic activity with high sensitive infrasound sensor using a piezoresistive cantilever. In *2019 IEEE 32nd International Conference on Micro Electro Mechanical Systems (MEMS)*, 783–786 (2019).
28. Grangeon, J. & Lesage, P. A robust, low-cost and well-calibrated infrasound sensor for volcano monitoring. *J. Volcanol. Geoth. Res.* **387**, 106668 (2019).
29. Schimmel, A. & Hübl, J. Automatic detection of debris flows and debris floods based on a combination of infrasound and seismic signals. *Landslides* **13**, 1181–1196 (2016).
30. Le Pichon, A. *et al.* Infrasound associated with 2004–2005 large sumatra earthquakes and tsunamis. *Geophys. Res. Lett.* **32**, 1 (2005).
31. Naderyan, V. *et al.* Mems microphone with 73dba snr in a 4mm x 3mm x 1.2 mm package. In *2021 21st International Conference on Solid-State Sensors, Actuators and Microsystems (Transducers)*, 242–245 (2021).
32. Liu, Y. *et al.* Epidermal mechano-acoustic sensing electronics for cardiovascular diagnostics and human–machine interfaces. *Sci. Adv.* **2**, e1601185 (2016).
33. Pasterkamp, H., Kraman, S. S. & Wodicka, G. R. Respiratory sounds: advances beyond the stethoscope. *Am. J. Respir. Crit. Care Med.* **156**, 974–987 (1997).
34. Gong, Z., Chen, K., Yang, Y., Zhou, X. & Yu, Q. Photoacoustic spectroscopy based multi-gas detection using high-sensitivity fiber-optic low-frequency acoustic sensor. *Sens. Actuators B Chem.* **260**, 357–363 (2018).
35. Luo, J. *et al.* Research on the detection of sf 6 decomposition products based on non-resonant photoacoustic spectroscopy. *Anal. Methods* **7**, 1200–1207 (2015).
36. Dehé, A., Wurzer, M., Fuldner, M. & Krumbein, J. Mems microphone with 73dba snr in a 4mm x 3mm x 1.2 mm package. In *AMA Conferences 2013*, 95–99 (2013).
37. Shin, K., Kim, C., Sung, M., Kim, J. & Moon, W. A modeling and feasibility study of a micro-machined microphone based on a field-effect transistor and an electret for a low-frequency microphone. *Sensors* **20**, 5554 (2020).
38. Rahaman, A., Park, C. H. & Kim, B. Design and characterization of a mems piezoelectric acoustic sensor with the enhanced signal-to-noise ratio. *Sens. Actuators A* **311**, 112087 (2020).
39. Chen, Y.-C. *et al.* On the PZT/SI unimorph cantilever design for the signal-to-noise ratio enhancement of piezoelectric mems microphone. *J. Micromech. Microeng.* **31**, 105003 (2021).
40. Takahashi, H., Dung, N. M., Matsumoto, K. & Shimoyama, I. Differential pressure sensor using a piezoresistive cantilever. *J. Micromech. Microeng.* **22**, 055015 (2012).
41. Wada, R. & Takahashi, H. Time response characteristics of a highly sensitive barometric pressure change sensor based on mems piezoresistive cantilevers. *Jpn. J. Appl. Phys.* **59**, 070906 (2020).
42. Nguyen, T.-V. *et al.* Highly sensitive low-frequency acoustic sensor using piezoresistive cantilever. In *2022 IEEE 35th International Conference on Micro Electro Mechanical Systems Conference (MEMS)*, 841–844 (2022).
43. Tortonese, M., Barrett, R. & Quate, C. Atomic resolution with an atomic force microscope using piezoresistive detection. *Appl. Phys. Lett.* **62**, 834–836 (1993).
44. Debbal, S. & Bereksi-Reguig, F. Time-frequency analysis of the first and the second heartbeat sounds. *Appl. Math. Comput.* **184**, 1041–1052 (2007).
45. Walker, H. K., Hall, W. D. & Hurst, J. W. *Clinical Methods: The History, Physical, and Laboratory Examinations* (Butterworths, 1990).

Acknowledgements

The photolithography masks were made using the University of Tokyo System Design Lab (d.lab)’s 8 inch EB writer F5112+VD01 donated by ADVANTEST Corporation. This study was supported by JSPS KAKENHI Grant No. 17H04903 and JPNP19005, subsidized by the New Energy and Industrial Technology Development Organization (NEDO).

Author contributions

Y.O., T.N., Y.T., H.O., and M.I. conceived the idea. Y.O. and T.N. fabricated the sensors, designed and conducted the experiments, and analyzed the data. H.T. conducted the low-frequency experiments.

Competing interests

The authors declare no competing interests.

Additional information

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